

Observer-based H_∞ Filtering of 2-D Singular System Described by Roesser Models¹⁾

XU Hui-Ling¹ LU Jun-Wei² ZOU Yun³

¹(Department of Applied Mathematics, Nanjing University of Science and Technology, Nanjing 210094)

²(School of Electrical and Automation Engineering, Nanjing Normal University, Nanjing 210042)

³(Department of Automation, Nanjing University of Science and Technology, Nanjing 210094)

(E-mail: xuhuilong-2005@yahoo.com.cn, we889@yahoo.com.cn, zouyun@vip.163.com)

Abstract This paper discusses the problem of the H_∞ filtering for discrete time 2-D singular Roesser models (2-D SRM). The purpose is to design an observer-based 2-D singular filter such that the error system is acceptable, jump modes free and stable, and satisfies a pre-specified H_∞ performance level. By general Riccati inequality and bilinear matrix inequalities (BMI), a sufficient condition for the solvability of the observer-based H_∞ filtering problem for 2-D SRM is given. A numerical example is provided to demonstrate the applicability of the proposed approach.

Key words 2-D singular systems, jump modes, general Riccati inequality, bilinear matrix inequalities

1 Introduction

In the past decades, 2-D singular systems have received much interest due to their extensive applications in many practical areas^[1~4]. An asymptotic stability theory based on the concept of jump modes was proposed in [1]. In [3] a singular observer design approach was developed while the problem of robust H_∞ control for uncertain 2-D singular Roesser models (2-D SRM) was considered in [2]. However, there is no remarkable progress to be reported on the problem of H_∞ filtering for 2-D singular systems. This motivates the present study.

In this paper, we consider the problem of observer-based H_∞ filtering for 2-D singular Roesser models (2-D SRM). Attention is focused on the design of 2-D singular filters such that the resulting closed-loop system is acceptable, jump modes free and stable, and satisfies a prespecified H_∞ performance level. General Riccati inequality approach and bilinear matrix inequalities(BMI) approach are presented for the design of observer-based H_∞ filters.

2 Problem formulation and preliminaries for 2-D singular systems

Consider the following 2-D SRM (Σ):

$$E \begin{bmatrix} \mathbf{x}^h(i+1, j) \\ \mathbf{x}^v(i, j+1) \end{bmatrix} = A \begin{bmatrix} \mathbf{x}^h(i, j) \\ \mathbf{x}^v(i, j) \end{bmatrix} + L\mathbf{d}(i, j) \quad (1)$$

$$\mathbf{y}(i, j) = C \begin{bmatrix} \mathbf{x}^h(i, j) \\ \mathbf{x}^v(i, j) \end{bmatrix} + D\mathbf{d}(i, j) \quad (2)$$

$$\mathbf{z}(i, j) = H \begin{bmatrix} \mathbf{x}^h(i, j) \\ \mathbf{x}^v(i, j) \end{bmatrix} \quad (3)$$

with the zero boundary conditions:

$$\mathbf{x}^h(0, j) = 0, \quad \mathbf{x}^v(i, 0) = 0 \quad (4)$$

where $\mathbf{x}^h(i, j) \in R^{n_1}$, $\mathbf{x}^v(i, j) \in R^{n_2}$ are the horizontal and vertical states, respectively, $\mathbf{y}(i, j) \in R^l$ is the measured output, $\mathbf{z}(i, j) \in R^p$ is the signal to be estimated, $\mathbf{d}(i, j) \in R^q$ is the noise signal which

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belongs to $l_2\{[0, \infty), [0, \infty)\}$. A, L, C, D , and H are real matrices of appropriate dimensions. $E \in R^{n \times n}$ is possibly singular, satisfying the 2-D regular pencil condition, *i.e.*, for some finite pair (z, w) :

$$\det[EI(z, w) - A] = \sum_{k=0}^{\bar{n}_1} \sum_{l=0}^{\bar{n}_2} a_{kl} z^k w^l \neq 0$$

where $I(z, w) = \text{diag}\{zI_{n_1}, wI_{n_2}\}$. When $a_{\bar{n}_1, \bar{n}_2} \neq 0$, system (Σ) is called acceptable.

In terms of the singular observer theory^[4], the following observer-based 2-D singular filter is adopted for studying the H_∞ filtering problem of 2-D SRM $(\hat{\Sigma})$:

$$E \begin{bmatrix} \hat{\mathbf{x}}^h(i+1, j) \\ \hat{\mathbf{x}}^v(i, j+1) \end{bmatrix} = A \begin{bmatrix} \hat{\mathbf{x}}^h(i, j) \\ \hat{\mathbf{x}}^v(i, j) \end{bmatrix} + K \left\{ \mathbf{y}(i, j) - C \begin{bmatrix} \hat{\mathbf{x}}^h(i, j) \\ \hat{\mathbf{x}}^v(i, j) \end{bmatrix} \right\} \quad (5)$$

$$\hat{\mathbf{z}}(i, j) = H \begin{bmatrix} \hat{\mathbf{x}}^h(i, j) \\ \hat{\mathbf{x}}^v(i, j) \end{bmatrix} \quad (6)$$

$$\hat{\mathbf{x}}^h(0, j) = 0, \quad \hat{\mathbf{x}}^v(i, 0) = 0 \quad (7)$$

Assumption 1. Systems (Σ) and $(\hat{\Sigma})$ are acceptable.

Remark 1. [3] presented that the assumption of acceptability is needed for the 2-D singular systems, because unacceptable systems are ill-posed in some ways.

Denote the state estimation error as

$$\mathbf{e}(i, j) = \begin{bmatrix} \mathbf{e}^h(i, j) \\ \mathbf{e}^v(i, j) \end{bmatrix} = \begin{bmatrix} \mathbf{x}^h(i, j) \\ \mathbf{x}^v(i, j) \end{bmatrix} - \begin{bmatrix} \hat{\mathbf{x}}^h(i, j) \\ \hat{\mathbf{x}}^v(i, j) \end{bmatrix}$$

and the estimation error for the signal z as

$$\mathbf{z}_e(i, j) = \mathbf{z}(i, j) - \hat{\mathbf{z}}(i, j)$$

Then, from (Σ) and $(\hat{\Sigma})$, we have the following dynamic error systems (Σ_e)

$$E \begin{bmatrix} \mathbf{e}^h(i, j) \\ \mathbf{e}^v(i, j) \end{bmatrix} = (A - KC) \begin{bmatrix} \mathbf{e}^h(i, j) \\ \mathbf{e}^v(i, j) \end{bmatrix} + (L - KD) \mathbf{d}(i, j) \quad (8)$$

$$\mathbf{z}_e(i, j) = H \begin{bmatrix} \mathbf{e}^h(i, j) \\ \mathbf{e}^v(i, j) \end{bmatrix} \quad (9)$$

and $G(z, w) = H[EI(zw) - A_k]^{-1}L_k$ is the transfer function matrix from the disturbances $\mathbf{d}(i, j)$ to the controlled output $\mathbf{z}_e(i, j)$, where $A_k = A - KC$, $L_k = L - KD$, and $\|G(z, w)\|_\infty = \sup_{\omega_1, \omega_2 \in [0, 2\pi]} \bar{\sigma}[G(e^{j\omega_1}, e^{j\omega_2})]$,

in which $\bar{\sigma}(\cdot)$ represents the maximum singular value of matrix (\cdot) .

The observer-based H_∞ filtering problem to be addressed in this paper can be formulated as follows: given a scalar $\gamma > 0$ and the 2-D SRM in (Σ) , find a 2-D singular filter of the form $(\hat{\Sigma})$, such that dynamic error systems (Σ_e) is acceptable, internally stable and jump modes free, and $\|G(z, w)\|_\infty < \gamma$.

Consider the following system of 2-D SRM (Σ_0)

$$E \begin{bmatrix} \mathbf{x}^h(i+1, j) \\ \mathbf{x}^v(i, j+1) \end{bmatrix} = A \begin{bmatrix} \mathbf{x}^h(i, j) \\ \mathbf{x}^v(i, j) \end{bmatrix} + L\mathbf{d}(i, j) \quad (10)$$

$$\mathbf{z}(i, j) = H \begin{bmatrix} \mathbf{x}^h(i, j) \\ \mathbf{x}^v(i, j) \end{bmatrix} \quad (11)$$

with zero boundary condition. We then have the following lemmas.

Lemma 1^[1]. 2-D SRM (Σ_0) is acceptable and internally stable if and only if

$$p(z, w) \neq 0, \quad 0 < |z| \leq 1, \quad 0 < |w| \leq 1 \quad (12)$$

where $p(z, w) = \det[E - AI(z, w)]$.

3 Observer-based 2-D singular H_∞ filter design

In this section, we shall present two approaches for the observer-based 2-D H_∞ filtering. They are the general Riccati inequality approach and bilinear matrix inequalities (BMI) approach.

Lemma 2. Given a positive scalar $\gamma > 0$, the 2-D SRM (Σ_0) with zero boundary condition is acceptable, internally stable and jump modes free, and satisfies $\|G(z, w)\|_\infty < \gamma$ if there exists a symmetric block-diagonal matrix $P = \text{diag}\{P_h, P_v\} \in R^{n \times n}$ such that the following LMIs hold

$$EPE^T \geq 0 \quad (13)$$

$$\begin{bmatrix} APA^T - EPE^T + LL^T & APH^T \\ HPA^T & HPH^T - \gamma^2 I \end{bmatrix} < 0 \quad (14)$$

$$\gamma^2 I - HPH^T > 0 \quad (15)$$

where $P_h \in R^{n_1 \times n_1}$ and $P_v \in R^{n_2 \times n_2}$.

Proof. From (14), it is easy to see that

$$APA^T - EPE^T < 0 \quad (16)$$

By this and (13), we assert that system (Σ_0) is acceptable, internally stable, and jump modes free. To show this, we suppose that there exist complex numbers z_1 and w_1 with $0 < |z_1| \leq 1, 0 < |w_1| \leq 1$ such that

$$p(z_1, w_1) = \det[E - AI(z_1, w_1)] = 0 \quad (17)$$

that is,

$$\det[E^T - I^H(z_1, w_1)A^T] = 0 \quad (18)$$

This implies that there exist some complex numbers z_0 and w_0 with $1 \leq |z_0| < \infty, 1 \leq |w_0| < \infty$ and a vector $x_0 \neq 0$ such that

$$(I(z_0, w_0)E^T - A^T)x_0 = 0 \quad (19)$$

By equation (19) one gets

$$x_0^H (APA^T - EPE^T)x_0 = (E^T x_0)^H \text{diag}\{(|z_0| - 1)P_h, (|w_0| - 1)P_v\}(E^T x_0) \geq 0 \quad (20)$$

This contradicts with (16). Hence, we have that 2-D SRM (Σ_0) is acceptable, stable internally, and jump modes free. Noting this and following a similar line as in the proof of Theorem 1 in [2], the desired result follows immediately. \square

3.1 General Riccati inequality approach

Assumption 2. $DD^T > 0$.

Note that Assumption 2 is standard in the Kalman and H_∞ filtering for 1-D systems. It implies that all the measurements are corrupted by noise.

Theorem 1. Consider the system (Σ) satisfying Assumption 2 and zero boundary condition. Given a prescribed level of H_∞ noise attenuation $\gamma > 0$, the H_∞ filtering problem is solvable if there exists a symmetric block-diagonal matrix $Q = \text{diag}\{Q_h, Q_v\} \in R^{n \times n}$, ($Q_h \in R^{n_1 \times n_1}, Q_v \in R^{n_2 \times n_2}$) such that

$$EQE^T \geq 0 \quad (21)$$

$$AQA^T - EQE^T - (AQ\bar{C}^T + L\bar{D}^T)(\bar{C}Q\bar{C}^T + \bar{R})^{-1}(\bar{C}QA^T + \bar{D}L^T) + LL^T < 0 \quad (22)$$

where $\bar{C} = \begin{bmatrix} C \\ H \end{bmatrix}$, $\bar{D} = \begin{bmatrix} D \\ 0 \end{bmatrix}$, $\bar{R} = \begin{bmatrix} DD^T & 0 \\ 0 & -\gamma^2 I \end{bmatrix}$.

In this situation, a suitable filter gain of (5) is given by

$$K = (AVC^T + LD^T)(CVC^T + DD^T) \quad (23)$$

where $V = Q + QH^T\theta^{-1}HQ$, $\theta = \gamma^2I - HQH^T > 0$.

Proof. From Lemma 2, the error system (Σ_e) is acceptable, internally stable and jump modes free, and satisfies $\|G(z, w)\|_\infty < \gamma$ if there exists a symmetric block-diagonal matrix $Q = \text{diag}\{Q_h, Q_v\} \in R^{n \times n}$, ($Q_h \in R^{n_1 \times n_1}, Q_v \in R^{n_2 \times n_2}$) such that

$$EQE^T \geq 0 \quad (24)$$

$$(A - KC)Q(A - KC)^T - EQE^T + (L - KD)(L - KD)^T + (A - KC)QH^T(\gamma^2I - HQH^T)^{-1}HQ(A - KC)^T < 0 \quad (25)$$

$$\theta = \gamma^2I - HQH^T > 0 \quad (26)$$

let $\Gamma_A = HQA^T$, $\Gamma_L = CQA^T + DL^T$, $\Gamma_C = CQA^T$

$$X_1 = DD^T + CQC^T + \Gamma_C^T\theta^{-1}\Gamma_C, \quad X_2 = \Gamma_L + \Gamma_C^T\theta^{-1}\Gamma_A \quad (27)$$

Then (25) can be rewritten as

$$AQA^T - EQE^T + LL^T + KX_1K^T - KX_2 - X_2^TK^T + \Gamma_A^T\theta^{-1}\Gamma_A < 0$$

In view of Assumption 2, $X_1 > 0$. Thus, it is easy to obtain

$$\begin{aligned} & AQA^T - EQE^T + LL^T - X_2^TX_1^{-1}X_2 + \Gamma_A^T\theta^{-1}\Gamma_A \\ & (-K^T + X_1^{-1}X_2)^TX_1(-K^T + X_1^{-1}X_2) < 0 \end{aligned} \quad (28)$$

From (22) and (27), it can be shown that $K = X_2^TX_1^{-1}$ and (22) is equivalent to

$$AQA^T - EQE^T + LL^T - X_2^TX_1^{-1}X_2 + \Gamma_A^T\theta^{-1}\Gamma_A < 0 \quad (29)$$

Hence, Q and K in (22) and (23) satisfy (29) or equivalently (29), *i.e.*, the H_∞ filtering problem is solvable. This completes the proof. \square

Remark 3. Solving a general Riccati inequality to obtain a block-diagonal solution may not be easy. In the following subsection, we shall propose a BMI approach to computing the filter gain.

3.2 BMI approach

Theorem 2. Consider the system (Σ) with zero boundary condition. Given a prescribed level of H_∞ noise attenuation $\gamma > 0$, the H_∞ filtering problem is solvable if there exist matrices $S \in R^{n \times n}$, $R_1 \in R^{n \times n}$, $R_2 \in R^{l \times n}$ and a symmetric block-diagonal matrix $P = \text{diag}\{P_h, P_v\} \in R^{n \times n}$, ($P_h \in R^{n_1 \times n_1}, P_v \in R^{n_2 \times n_2}$) such that

$$EPE^T \geq 0 \quad (30)$$

$$\gamma^2I - HPH^T > 0 \quad (31)$$

$$\begin{bmatrix} -EPE^T + A_kR_1^T + R_1A_k^T & A_kR_2^T + R_1H^T & -R_1 + A_kS & L_k \\ R_2A_k^T + HR_1^T & -\gamma^2I + HR_2^T + R_2H^T & -R_2 + HS & 0 \\ -R_1^T + S^TA_k^T & -R_2^T + S^TH^T & P - S - S^T & 0 \\ L_k^T & 0 & 0 & -I \end{bmatrix} < 0 \quad (32)$$

where $A_k = A - KC$, $L_k = L - KD$.

Proof. By the Schur complement formula, (32) equivalent to

$$\begin{bmatrix} -EPE^T + LL^T + A_kR_1^T + R_1A_k^T & A_kR_2^T + R_1H^T & -R_1 + A_kS \\ R_2A_k^T + HR_1^T & -\gamma^2I + HR_2^T + R_2H^T & -R_2 + HS \\ -R_1^T + S^TA_k^T & -R_2^T + S^TH^T & P - S - S^T \end{bmatrix} < 0 \quad (33)$$

Let $W = \begin{bmatrix} -EPE^T + LL^T & 0 \\ 0 & -\gamma^2I \end{bmatrix}$ and $\Psi = [A_k^T \quad H^T]$, $R = \begin{bmatrix} R_1 \\ R_2 \end{bmatrix}$

Then (32) can be written as

$$\begin{bmatrix} W + \Psi^TR^T + R\Psi & -R + \Psi^TS \\ -R^T + S^T\Psi & P - S - S^T \end{bmatrix} < 0 \quad (34)$$

Note that this is in turn equivalent to^[4] $W + \Psi^T P \Psi < 0$

Then

$$W + \Psi^T P \Psi = \begin{bmatrix} -EPE^T + LL^T + A_k P A_k^T & A_k P H^T \\ H P A_k^T & -\gamma^2 I + H P H^T \end{bmatrix} < 0 \quad (35)$$

Therefore, the desired result follows immediately from (30), (31) and Lemma 1. This completes the proof. \square

From Theorem 2, the following iterative algorithm can be used to solve the uncertain 2-D SRM H_∞ filtering problem.

4 Algorithm

Step 1. Choose an initial K , and solve the following convex optimization problem:

$$\begin{aligned} & \min_{(P,S,R)} \{\mu\} \\ \text{Such that } & \begin{bmatrix} W + \Psi^T R^T + R \Psi & -R + \Psi^T S \\ -R^T + S^T \Psi & P - S - S^T \end{bmatrix} < \mu I \end{aligned}$$

and (29)~(30) are hold. If $\mu \leq 0$, then the problem is solved; otherwise, go to Step 2.

Step 2. With the obtained matrices P, S , and R , solve the above optimization with respect to K . Again, if $\mu \leq 0$, the problem is solved; otherwise, go to Step 1.

5 Numerical example

Consider a 2-D SRM (Σ) with parameters: ($n_1 = 1, n_2 = 2$)

$$E = \begin{bmatrix} 1 & 0 & 1 \\ 0 & 1 & 0 \\ 0 & 0 & 0 \end{bmatrix}, A = \begin{bmatrix} 2 & 1 & 0 \\ 0 & 0 & 1 \\ 0 & 0 & 1 \end{bmatrix}, C = [1 \quad 0.2 \quad 1], L = \begin{bmatrix} 0.1 \\ 0.2 \\ 0.3 \end{bmatrix}, H = [0.1 \quad 0.1 \quad 0.2], D = 0.2$$

It is easy to see that this system is an acceptable, unstable system with jump-mode.

Let $\gamma = 0.5$. By the above algorithm, the solution to BMIs (29)~(31) is as follows.

$$\begin{aligned} P &= \begin{bmatrix} 6.1293 & 0 & 0 \\ 0 & 2.6159 & -2.2767 \\ 0 & -2.2767 & -2.1312 \end{bmatrix}, R_1 = \begin{bmatrix} -1050.7 & 878.3 & 749.0 \\ 878.3 & 734.1 & -625.5 \\ 749.0 & -625.5 & -533.9 \end{bmatrix} \\ R_2 &= [159.8490 \quad -133.4886 \quad -113.5674] \end{aligned}$$

The corresponding filter gain can be obtained as: $K = [2.7718 \quad -0.1603 \quad -1.0000]^T$

Fig. 1 shows the frequency response of the error system (Σ_e) over all frequencies. It can be observed that the amplitude response of the filtering error transfer function is below the prescribed H_∞ noise attenuation level $\gamma = 0.5$.

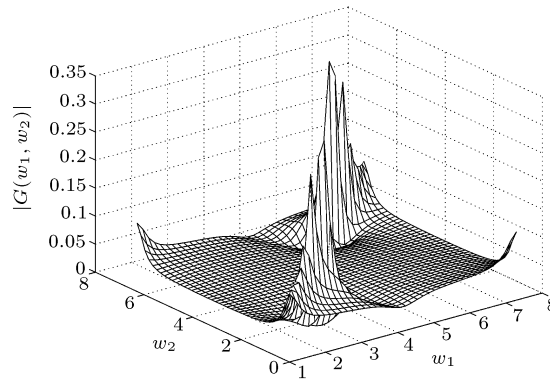


Fig. 1 Frequency response of the filtering error system

6 Conclusions

This paper has solved the H_∞ filtering problem for 2-D singular Roesser models. Both the general Riccati inequality and bilinear matrix inequalities (BMI) have been developed for the design of an observer-based 2-D singular H_∞ filter. Numerical example is provided to demonstrate the applicability of the proposed approach.

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XU Hui-Ling Received her Ph. D. degree from Nanjing University of Science and Technology in 2005. Her research interests include singular system, 2-D systems, and robust control.

LU Jun-Wei Received her bachelor degree from the Nanjing University of Aeronautics and Astronautics, Nanjing, P. R. China in 2001. Her research interests include robust filtering and control, time-delay systems, and nonlinear systems.

ZOU Yun Received his Ph. D. degree in automation from Nanjing University of Science and Technology in 1990. His research interests include singular systems, multidimensional systems, and transient stability of power systems.